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Practical Digital Pre-Distortion Techniques for PA Linearization in 3GPP LTE

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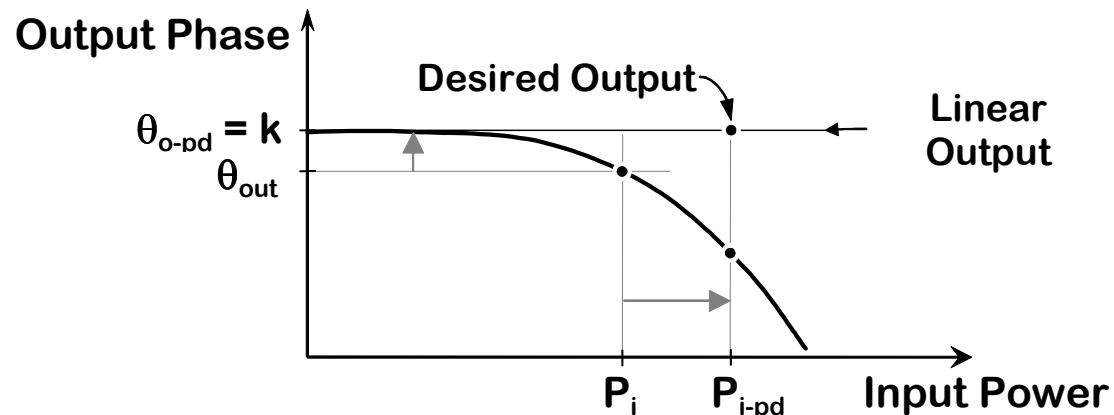
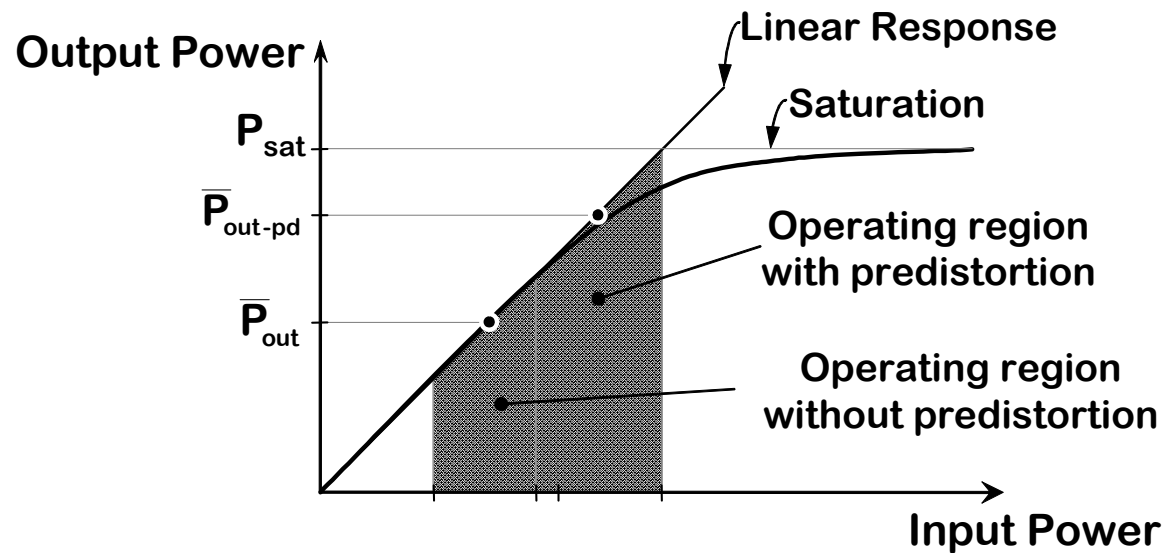


Agenda

- Digital PreDistortion----Principle
- Crest Factor Reduction
- Digital PreDistortion Simulation
- Digital PreDistortion Hardware Verification



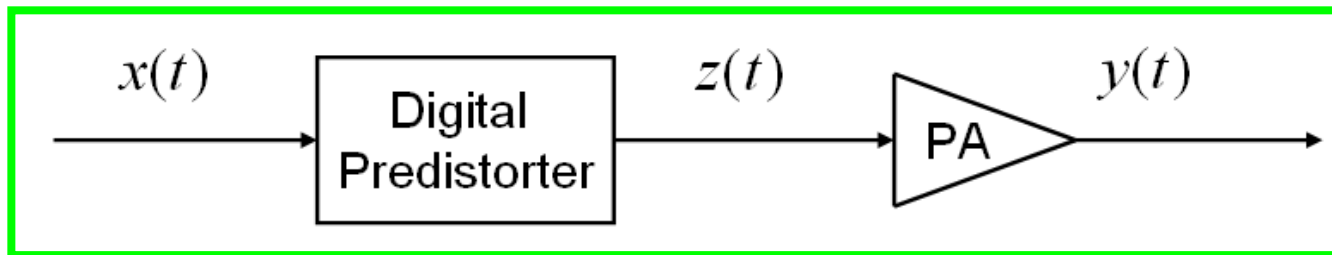
Digital Pre-Distortion----- Principle



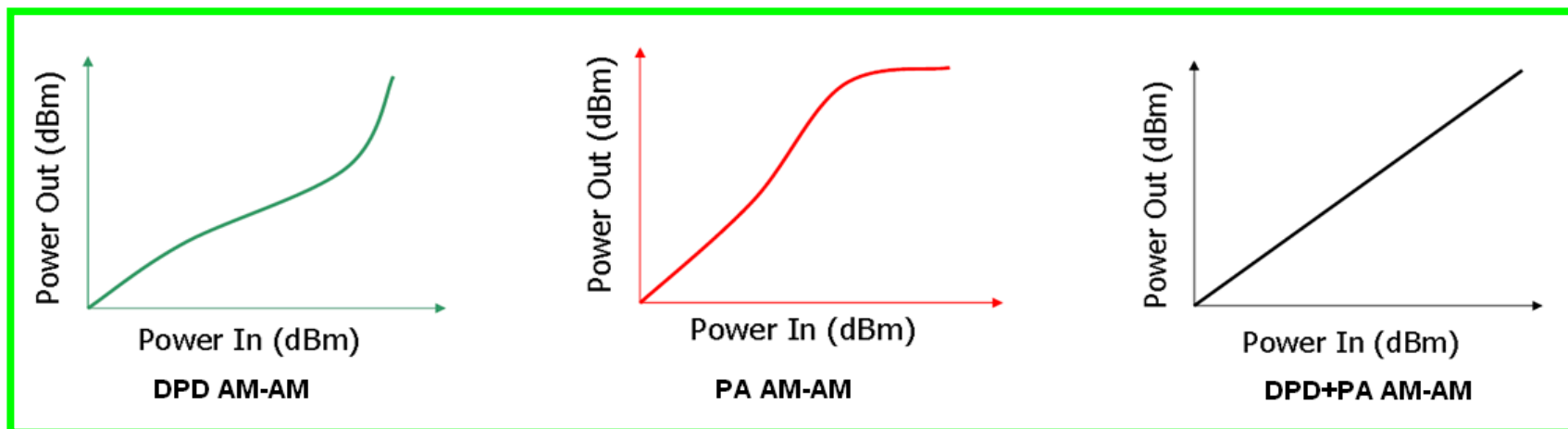
Digital Pre-Distortion----- Principle

The DPD-PA cascade attempts to combine two nonlinear systems into one linear result which allows the PA to operate closer to saturation.

The objective of digital predistorter is to have $y(t) \approx Cx(t)$, where C is a constant.



The most important step is to extract PA nonlinear behavior accurately and efficiently.



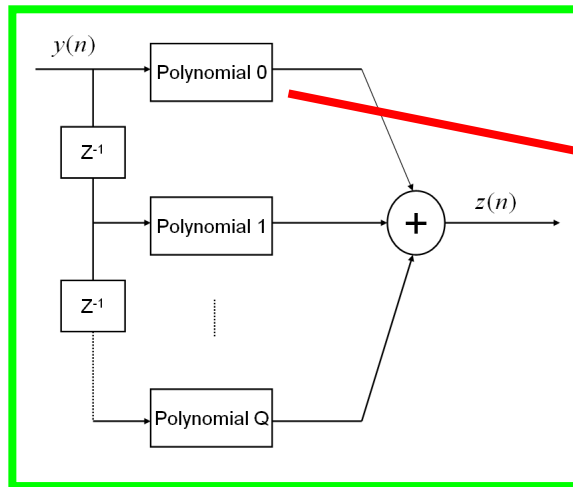
Memory Polynomial Algorithm

- As the signal (such as 3GPP LTE) bandwidth gets wider, power amplifiers begin to exhibit memory effects. Memoryless (LUT) pre-distortion can achieve only very limited linearization performance.
- Volterra series is a general nonlinear model with memory. It is unattractive for practical applications because of its large number of coefficients.
- Memory polynomial reduces Volterra's model complexity. It is interpreted as a special case of a generalized Hammerstein model. Its equation is as follows:

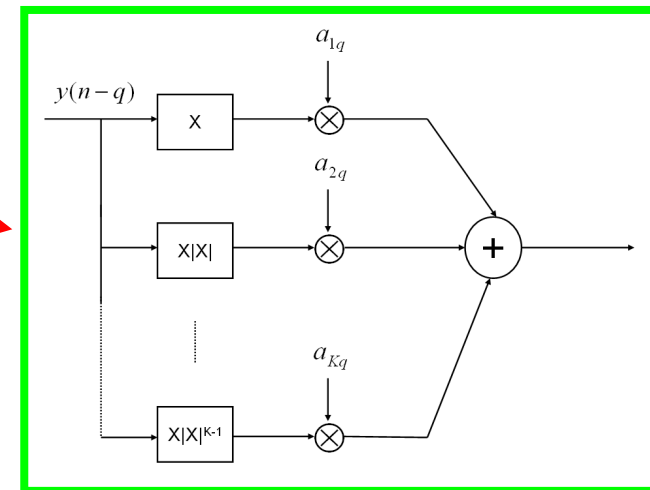
$$z(n) = \sum_{k=1}^K \sum_{q=0}^Q a_{kq} y(n-q) |y(n-q)|^{k-1}$$

K is Nonlinearity order and Q is Memory order

Memory Polynomial Structure

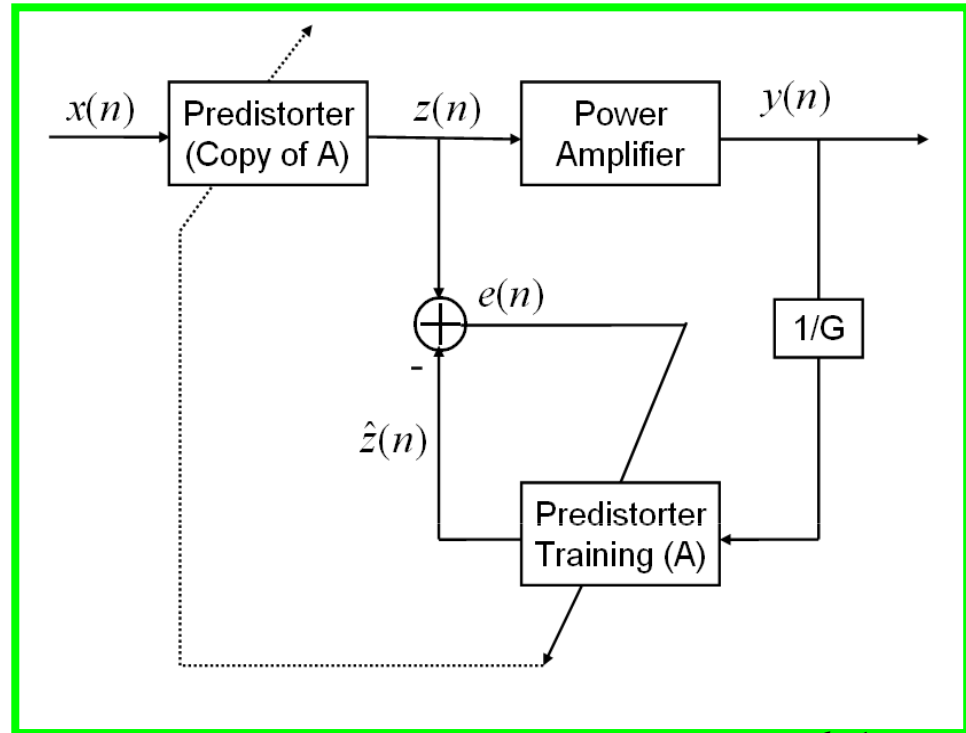


Polynomial Structure



Signal Training to derive the Memory Polynomial

1. Pre-distorter training: Nonlinear coefficients are extracted from the PA input and PA output waveforms (ie – on real physical behavior)
2. Copy of PA: The DPD model accurately captures the nonlinearity with memory effects



Memory Polynomial Coefficients

$$\hat{\mathbf{a}} = (\mathbf{U}^H \mathbf{U})^{-1} \mathbf{U}^H \mathbf{z}$$

$$\hat{\mathbf{a}} = [\hat{a}_{10}, \dots, \hat{a}_{K0}, \dots, \hat{a}_{1Q}, \dots, \hat{a}_{KQ}]^T$$

$$\mathbf{z} = [z(0), z(1), \dots, z(N-1)]^T$$

$$u_{kq}(n) = \frac{y(n-q)}{G} \left| \frac{y(n-q)}{G} \right|^{k-1}$$

$$\mathbf{u}_{kq} = [u_{kq}(0), u_{kq}(1), \dots, u_{kq}(N-1)]^T$$

$$\mathbf{U} = [\mathbf{u}_{10}, \dots, \mathbf{u}_{K0}, \dots, \mathbf{u}_{1Q}, \dots, \mathbf{u}_{KQ}]$$

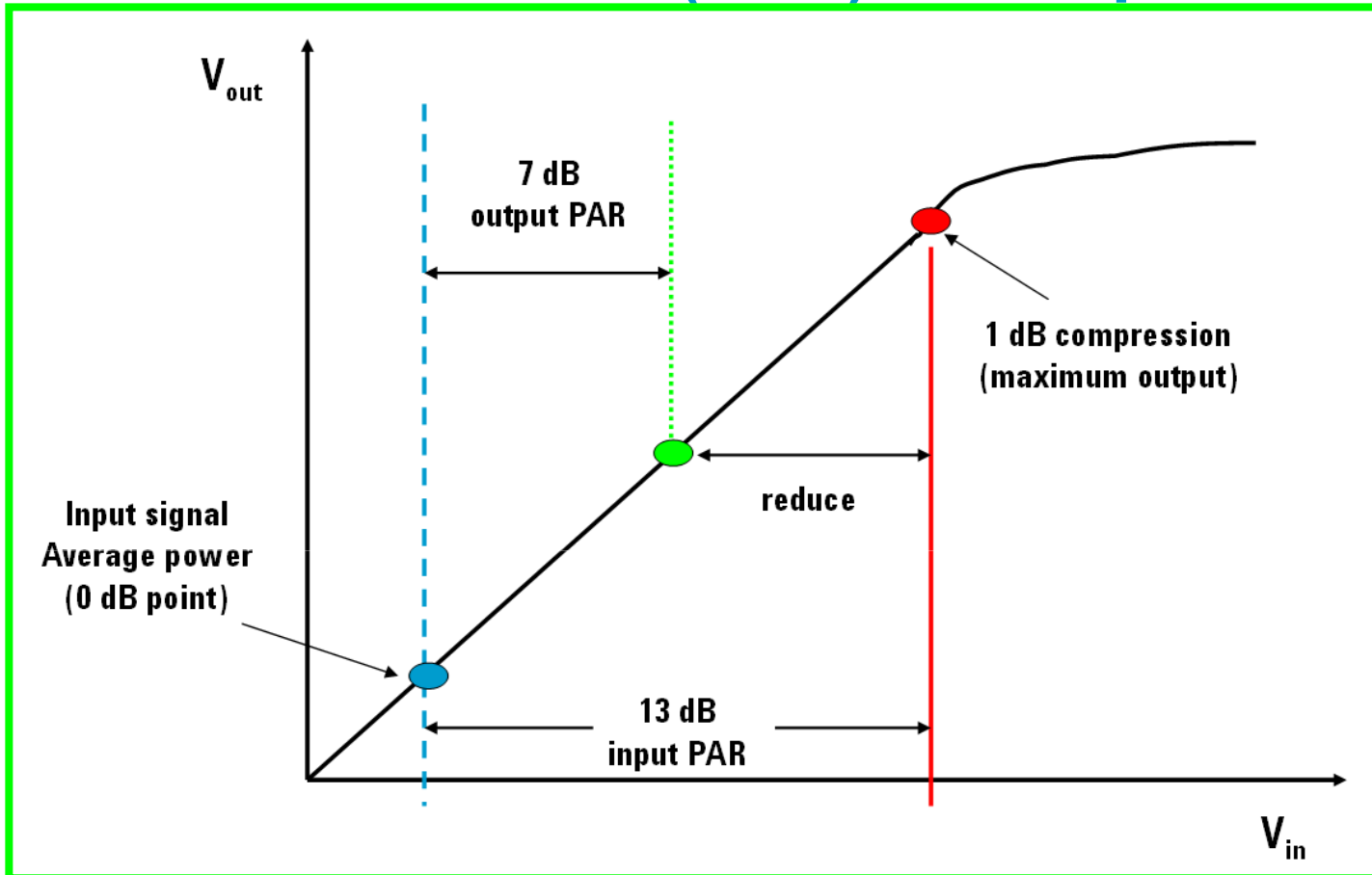
Crest Factor Reduction (CFR) Concepts

- Spectrally efficient wideband RF signals may have PAPR >13dB.
- CFR preconditions the signal to reduce signal peaks without significant signal distortion
- CFR allows the PA to operate more efficiently – it is not a linearization technique
- CFR supplements DPD and improves DPD effectiveness
- Without CFR and DPD, a basestation PA must operate at significant back-off from saturated power to maintain linearity. The back-off reduces efficiency

Benefits of CFR

1. PAs can operate closer to saturation, for improved efficiency (PAE).
2. Output signal still complies with spectral mask and EVM specifications

Crest Factor Reduction (CFR) Concepts

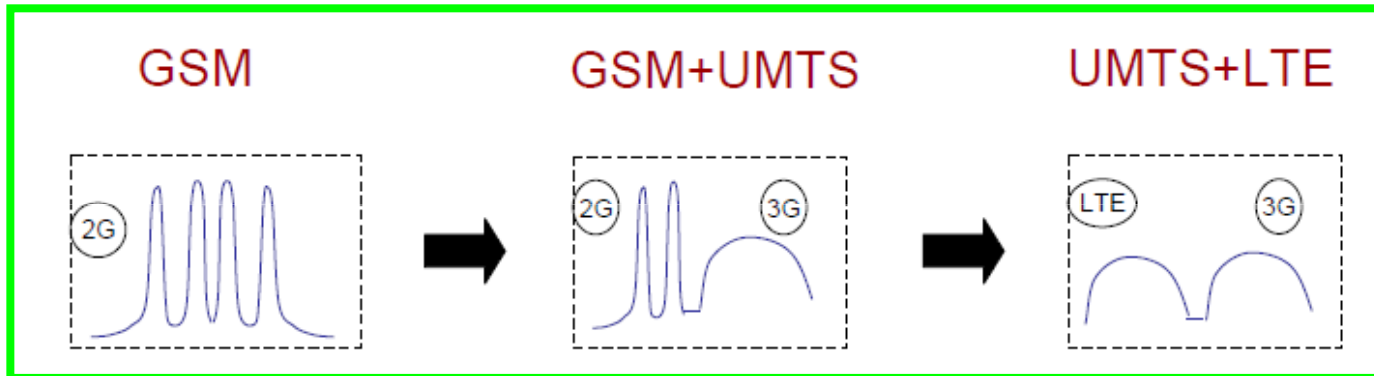


If you can reduce the Peak-to-Average Ratio of the signal, then for a given amplitude Peak, you can raise the Average power (up & to the right, above) with no loss in signal quality.

Thus, CFR enables higher PA efficiency by reducing the back-off, often by 6dB

Crest Factor Reduction for Multiple-Carrier Signals

- Multiple-Carrier Signals (such as GSM, WCDMA, WiMAX) already have high PAPR.
- In the future, they will also include multiple waveforms (ie - LTE with 3G WCDMA).
- Therefore CFR will increase in importance for Multi-Carrier PA (MCPA) linearization.

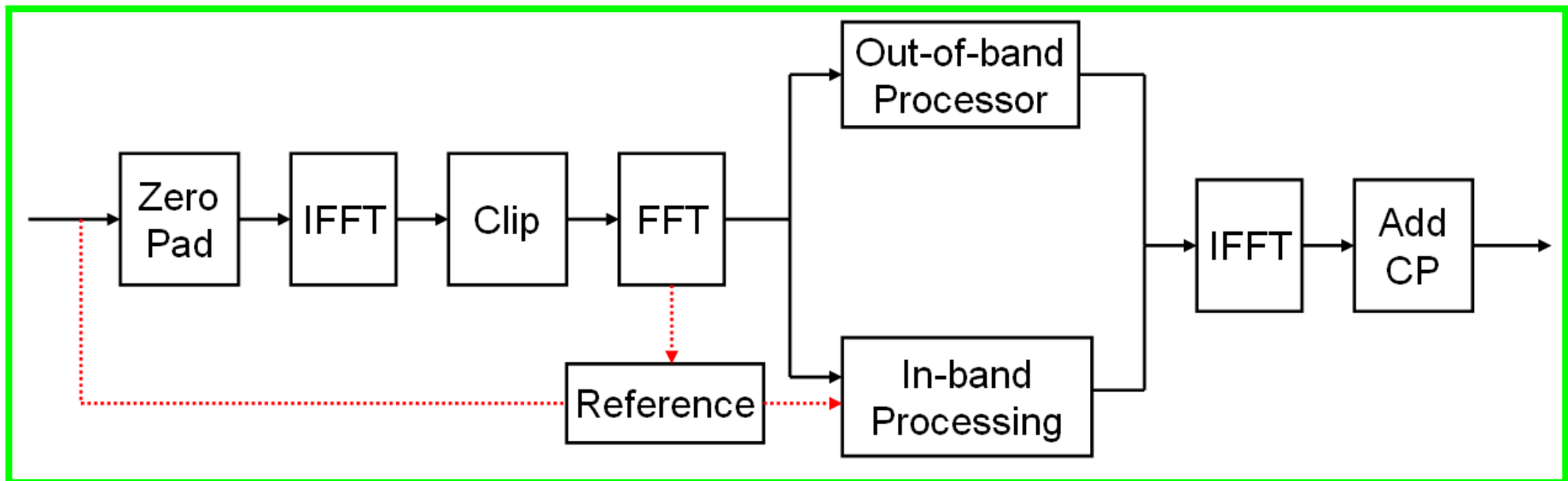


CFR algorithm for multiple carrier signals

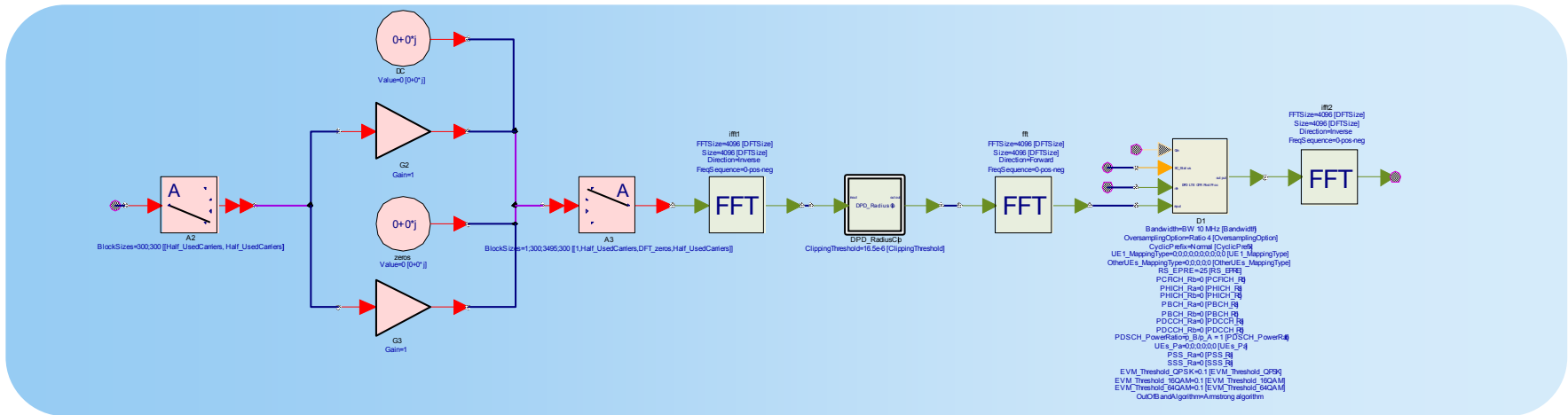
- PW (Peak Windowing)-CFR
- NS (Noise-Shaping) -CFR
- PI (Pulse Injection)-CFR
- PC (Peak Cancellation)-CFR

CFR for 3GPP LTE DL OFDM Signal

- Controls EVM and band limits in the frequency domain.
 - Constrains constellation errors, to avoid bit errors.
 - Constrains the degradation on individual sub-carriers.
- Allows QPSK sub-carriers to be degraded more than 64 QAM sub-carriers.
- Does not degrade reference signals, P-SS and S-SS.
- All control channels (PDCCH, PBCH, PCFICH and PHICH) adopts QPSK threshold.



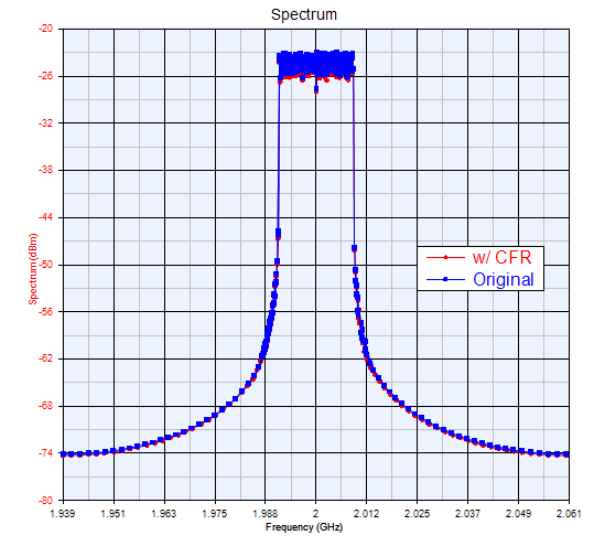
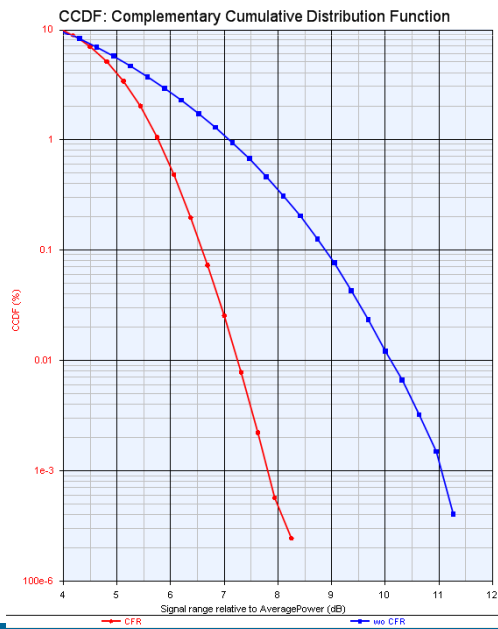
LTE CFR (Crest Factor Reduction)



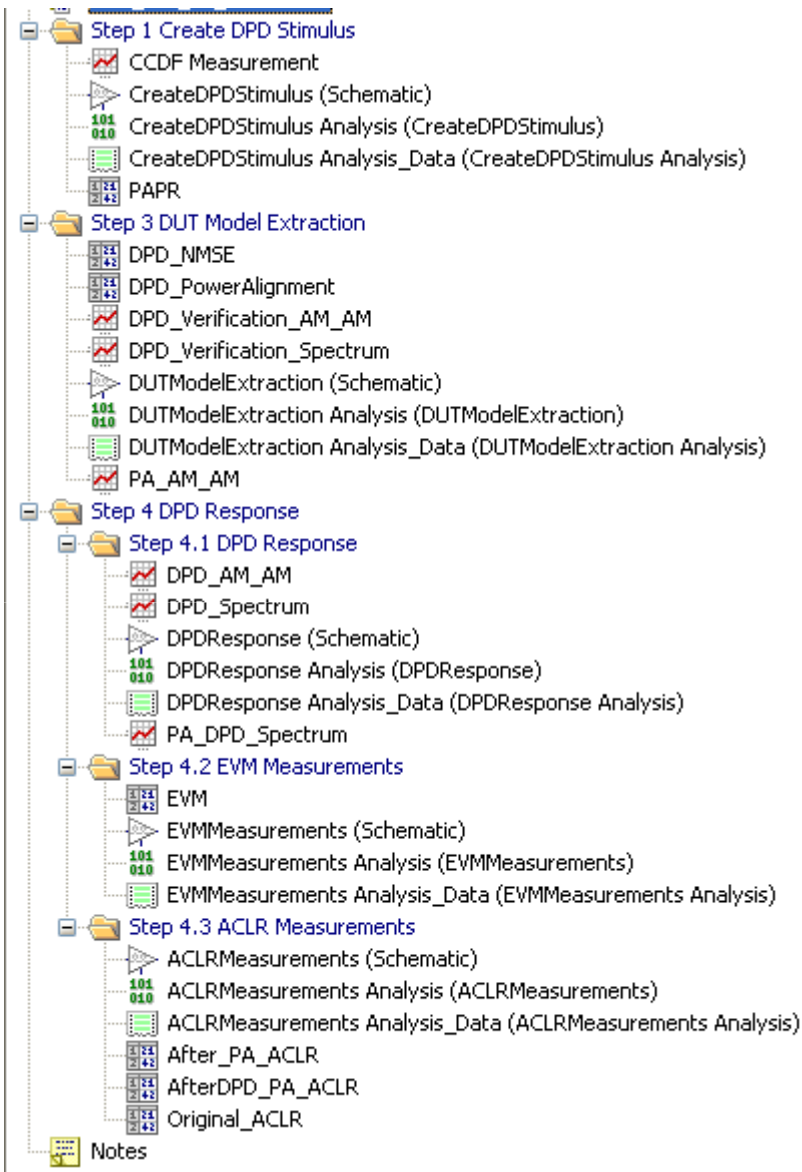
Simulation Results

LTE Downlink 10MHz,
 Sampling Rate 61.44MHz,
 QPSK,
 EVM threshold 10%

| Origin_PAPR | CFR_PAPR |
|-------------|----------|
| 9.05 | 6.685 |



DPD Simulation Workspace



Step 1 is to Generate Waveform for DPD

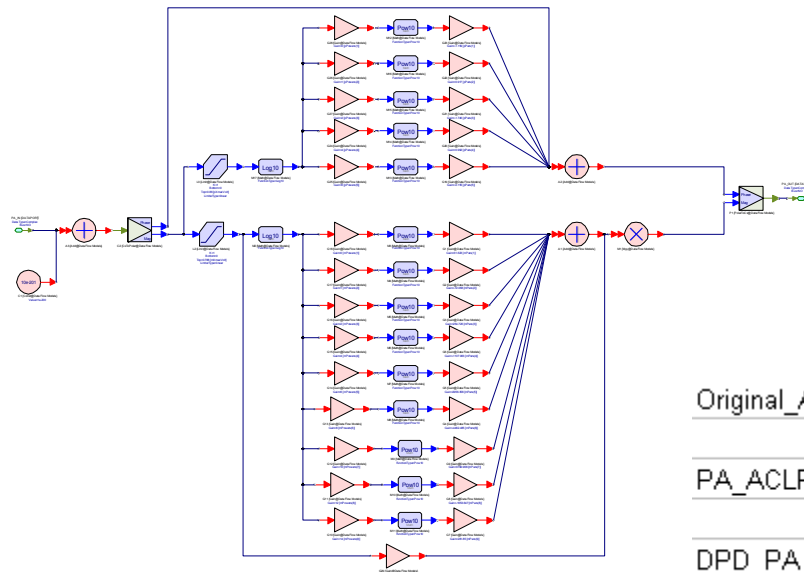
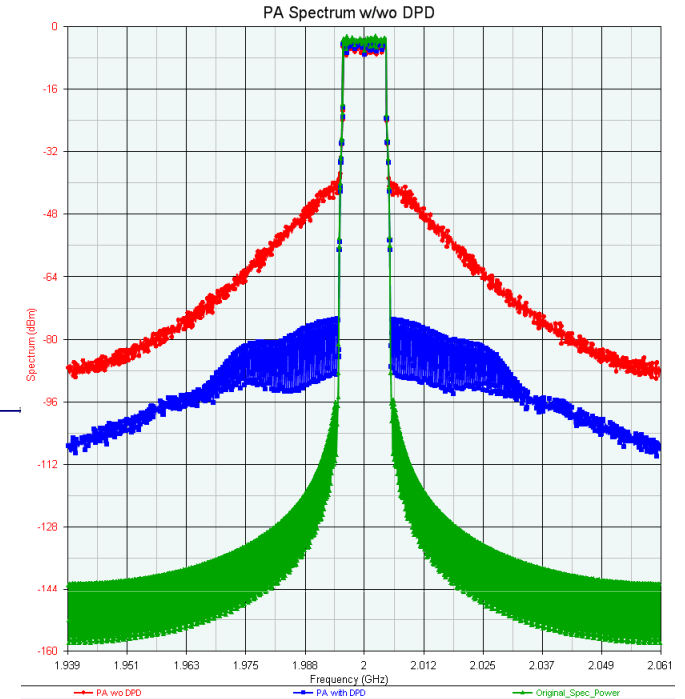
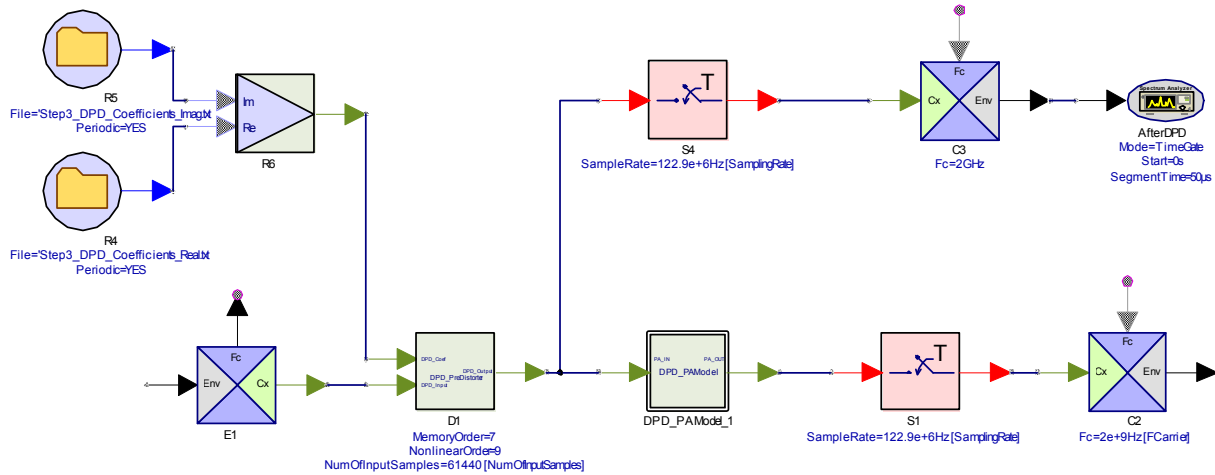
Step 3 is for DUT Model Extraction

Step 4 is for DPD Response

Compared with hardware verification tool, simulation tool does not include Step 2 and Step 5.

Hardware verification tool will be introduced later.

LTE DPD simulation for a memoryless nonlinear PA



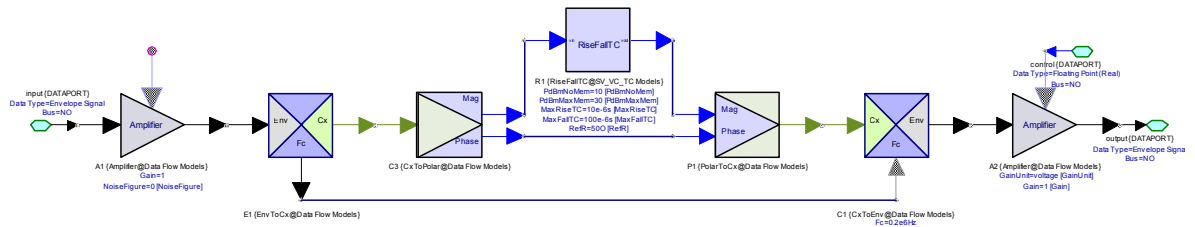
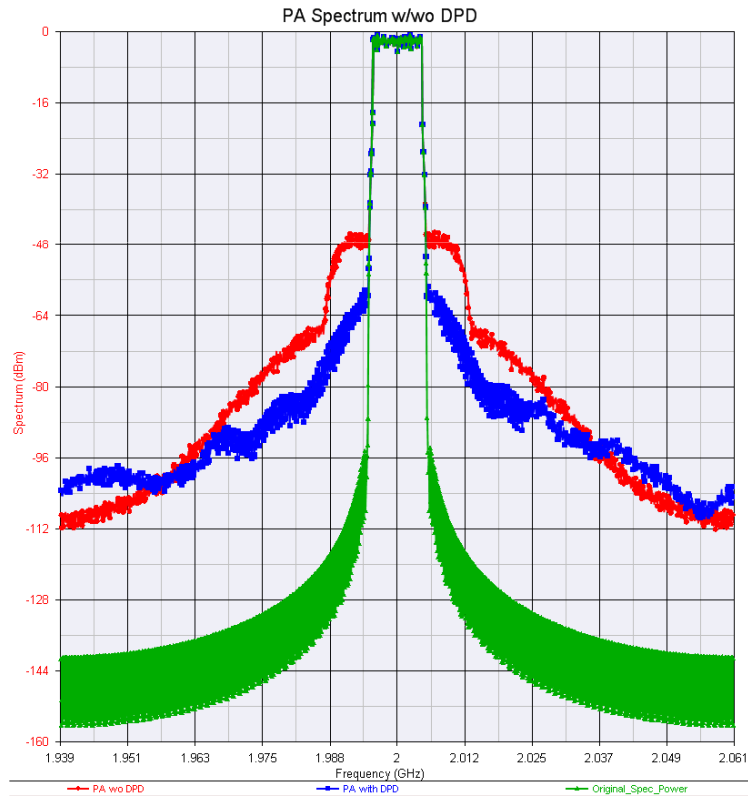
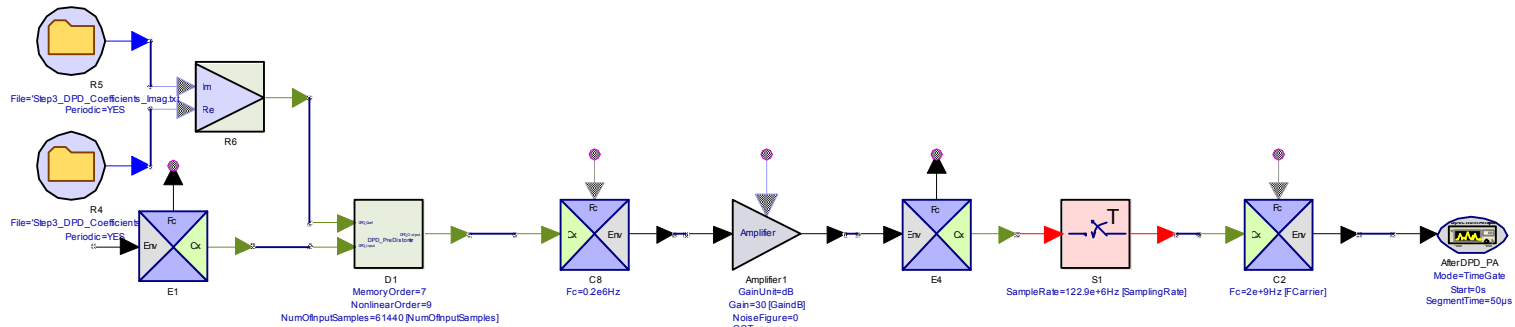
EVM (dB)

| Index | AfterDPD_PA | Original | After_PA |
|-------|-------------|----------|----------|
| 1 | -26.802 | -26.804 | -24.986 |
| 2 | -26.787 | -26.789 | -24.974 |

ACLR (dB)

| Original_ACLR_L_2BW | Original_ACLR_L_BW | Original_ACLR_U_BW | Original_ACLR_U_2BW |
|---------------------|--------------------|--------------------|---------------------|
| 72.036 | 64.23 | 63.542 | 71.869 |
| PA_ACLR_L_2BW | PA_ACLR_L_BW | PA_ACLR_U_BW | PA_ACLR_U_2BW |
| 51.075 | 38.899 | 38.963 | 51.151 |
| DPD_PA_ACLR_L_2BW | DPD_PA_ACLR_L_BW | DPD_PA_ACLR_U_BW | DPD_PA_ACLR_U_2BW |
| 71.651 | 64.067 | 63.404 | 71.497 |

LTE DPD simulation for a nonlinear PA with memory



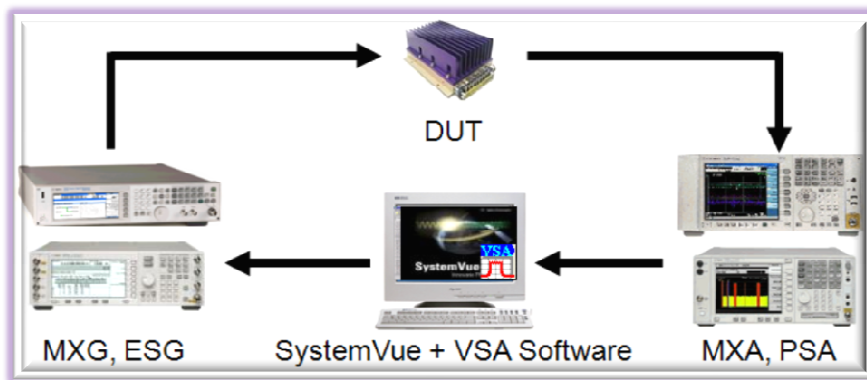
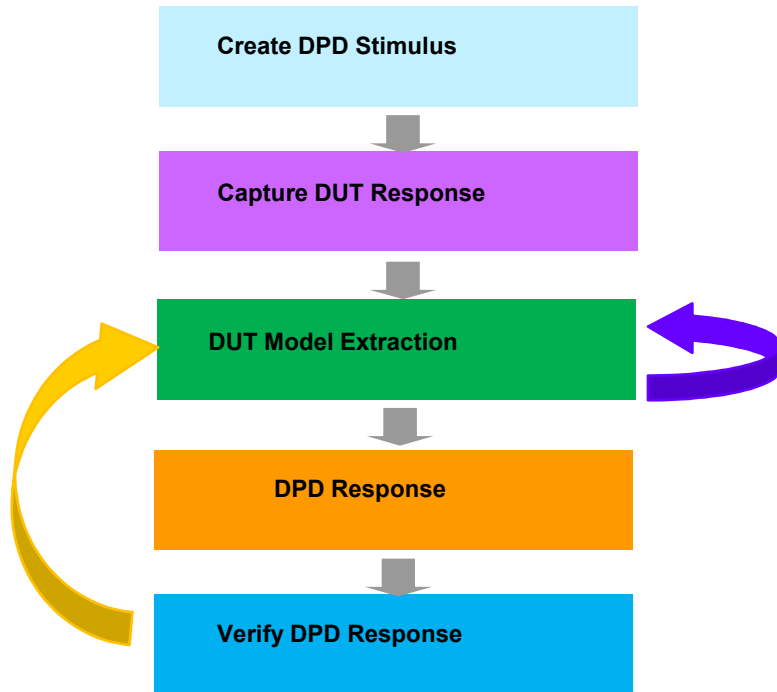
EVM (dB)

| Index | AfterDPD_PA | Original | After_PA |
|-------|-------------|----------|----------|
| 1 | -26.792 | -26.804 | -26.673 |
| 2 | -26.777 | -26.789 | -26.659 |

ACLR (dB)

| Original_ACLR_L_2BW | Original_ACLR_L_BW | Original_ACLR_U_BW | Original_ACLR_U_2BW |
|---------------------|--------------------|--------------------|---------------------|
| 72.036 | 64.23 | 63.542 | 71.869 |
| PA_ACLR_L_2BW | PA_ACLR_L_BW | PA_ACLR_U_BW | PA_ACLR_U_2BW |
| 67.729 | 47.099 | 47.133 | 67.713 |
| DPD_PA_ACLR_L_2BW | DPD_PA_ACLR_L_BW | DPD_PA_ACLR_U_BW | DPD_PA_ACLR_U_2BW |
| 71.841 | 61.013 | 60.504 | 71.871 |

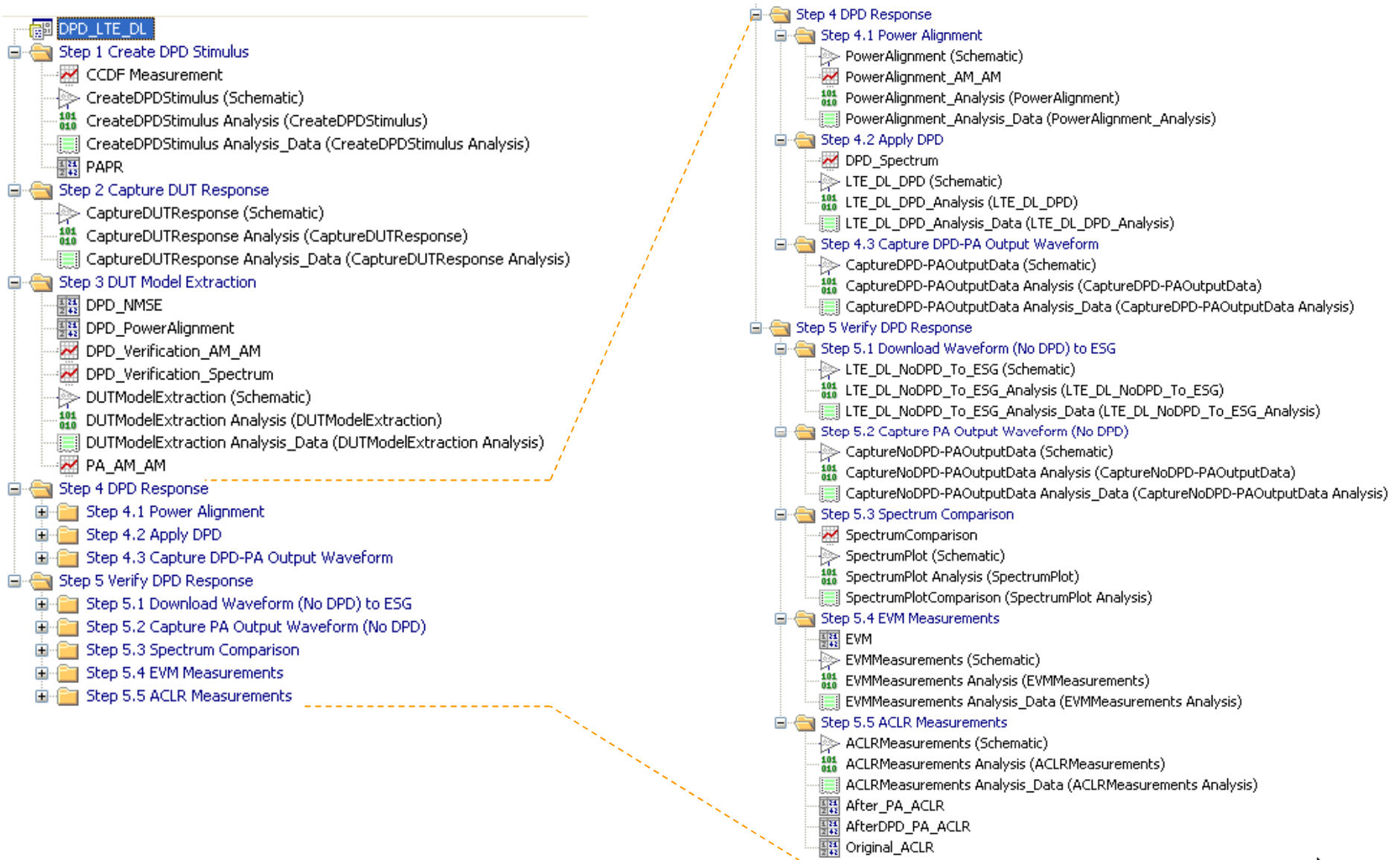
DPD Hardware Verification Flowchart



DPD HW Flowchart consists of 5 steps:

- Step 1 (Create DPD Stimulus) is to download waveform (LTE or User defined) into ESG/MXG.
- Step 2 (Capture DUT Response) is to capture both waveforms before power amplifier and after power amplifier from PSA/MXA/PXA by using VSA89600 software.
- Step 3 (DUT Model Extraction) is to extract PA nonlinear coefficients based on both captured PA input and PA output waveforms and then to verify DPD by using PA nonlinear coefficients.
- Step 4 (DPD Response) is to download the waveform (LTE or User Defined) after pre-distorter (by using PA nonlinear coefficient from Step 3) into ESG/MXG, this real signal passes through the PA DUT, capture PA output waveform from PSA/MXA/PXA by using VSA89600 software.
- Step 5 (Verify DPD Response) is to show the performance improvement after DPD.

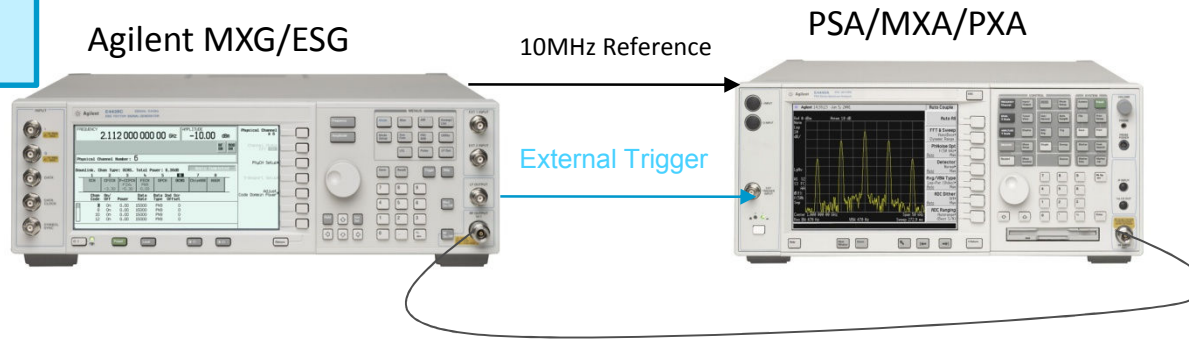
DPD Hardware Verification Workspace Structure



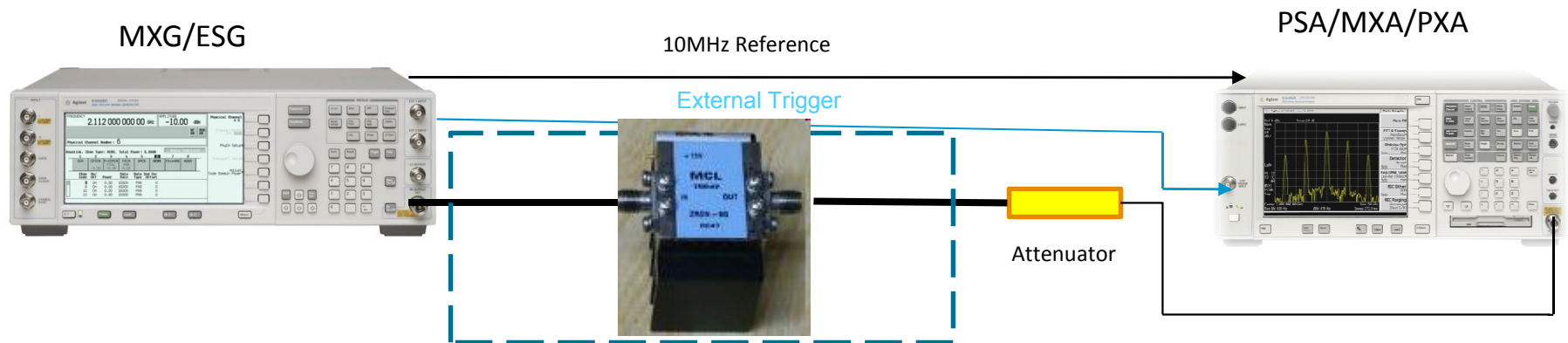
DPD Hardware Verification Platform

1. PA input signal capture

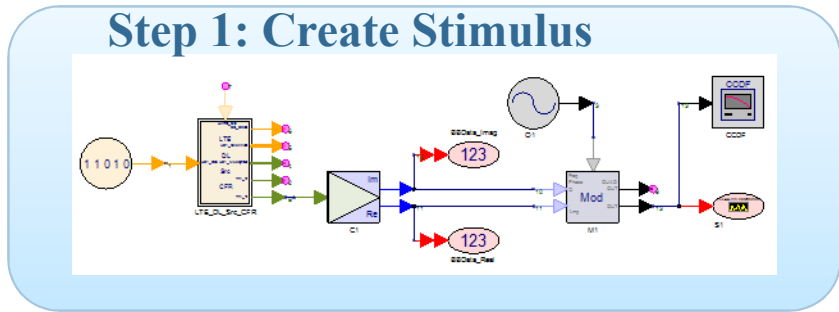
Signal source:
LTE 10MHz



2. PA output signal capture



DPD Hardware Verification – LTE (Step 1)



The CFR must be enable in LTE source.
 LTE paramters (such as bandwidth, Resource Block allocation and etc) can be set.

The download waveform transmit power, length also can be set.

The screenshot shows the SystemVue DPD software interface. The main window is titled 'DPD' and has five tabs: '1: Create DPD Stimulus', '2: Capture DUT Response', '3: DUT Model Extraction', '4: DPD Response', and '5: Verify DPD response'. The '1: Create DPD Stimulus' tab is active.

LTE Parameters:

- FCarrier: 2.0e9 Hz
- Bandwidth: 3: BW 10 MHz (50 RB)
- Oversampling Option: Ratio 4
- Cyclic Prefix (CP) Type: Normal (Nsc: 12 | Nsymb: 7)
- CFREnable:
- Clipping Threshold: 0.0000165
- QPSK_EVMThreshold: 0.1
- 16QAM_EVMThreshold: 0.1
- 64QAM_EVMThreshold: 0.1

Mapping Type (Modulation Scheme):

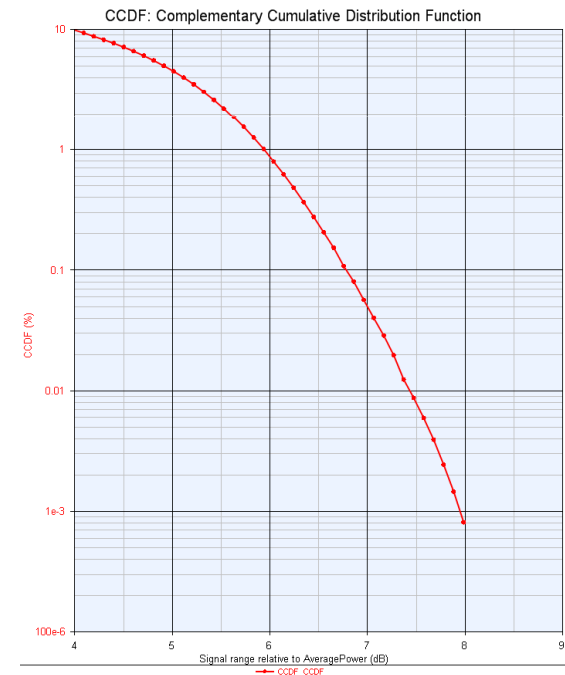
Subframe 0 to 9: CW QPSK QPSK QPSK QPSK QPSK QPSK QPSK QPSK QPSK

Buttons: Edit User Mapping, Current UE: 1, Prev UE, Next UE

Download Parameters:

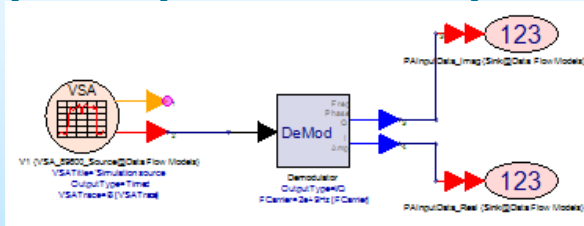
- RFPower: -2 dBm
- PrimAddress: 146.208.175.2
- TimeStart: 0 ms
- TimeStop: 30 ms

Buttons: Download Waveform, Go To ESG Web Control, CCDF, PAPR: 5.85698777



DPD Hardware Verification – LTE (Step 2)

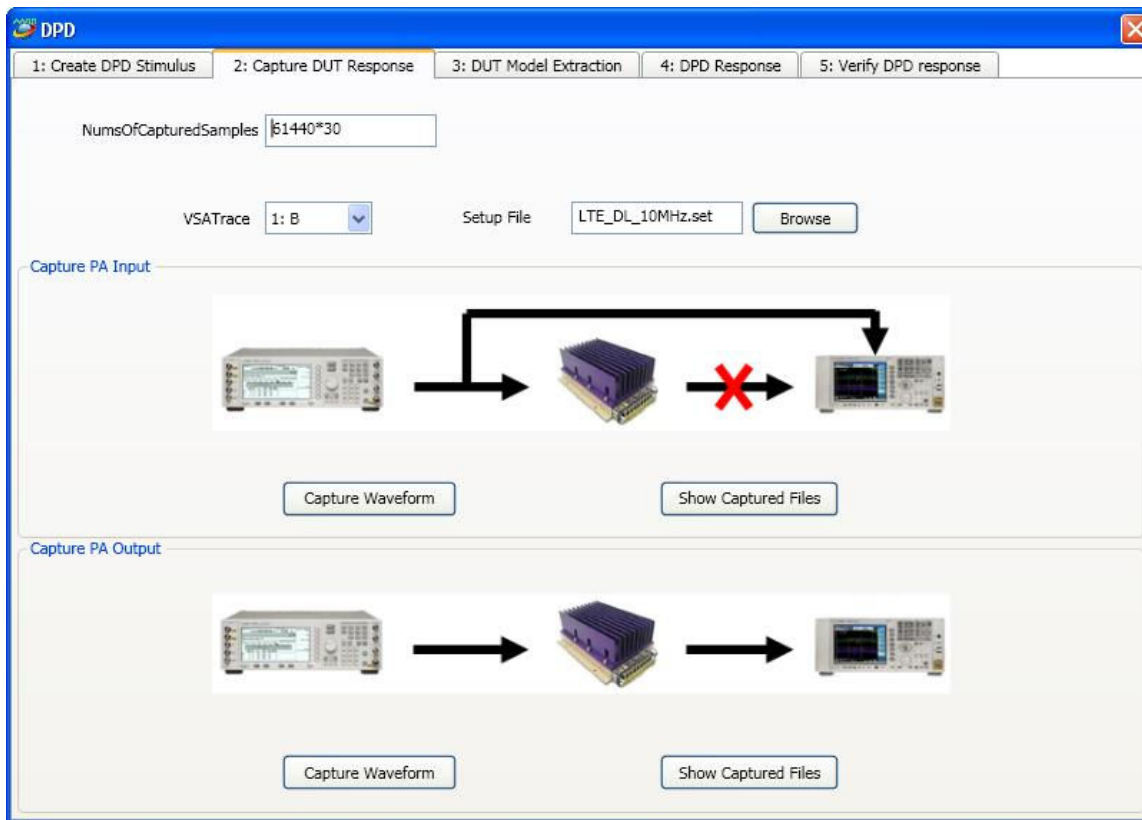
Step 2: Capture DUT Response



Firstly, connect the ESG directly with the PSA/PXA and click the “Capture Waveform” button in the “Capture PA Input” panel in the GUI. The captured signal is the input of the PA DUT.

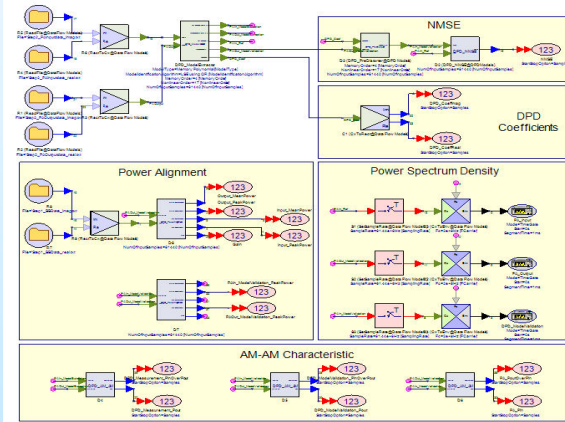
Then, connect the ESG with the DUT, and then connect the DUT with the PSA/PXA and click the “Capture Waveform” button in the “Capture PA Output” panel in the GUI. The captured signal is the output of the PA DUT.

These I/Q files are stored for further usage.

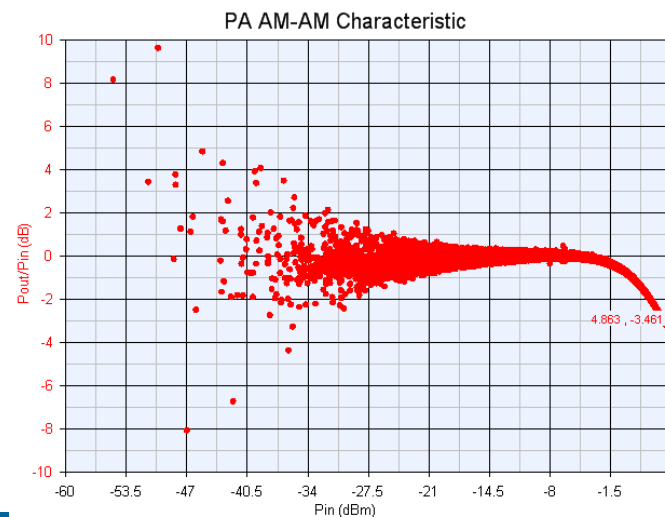
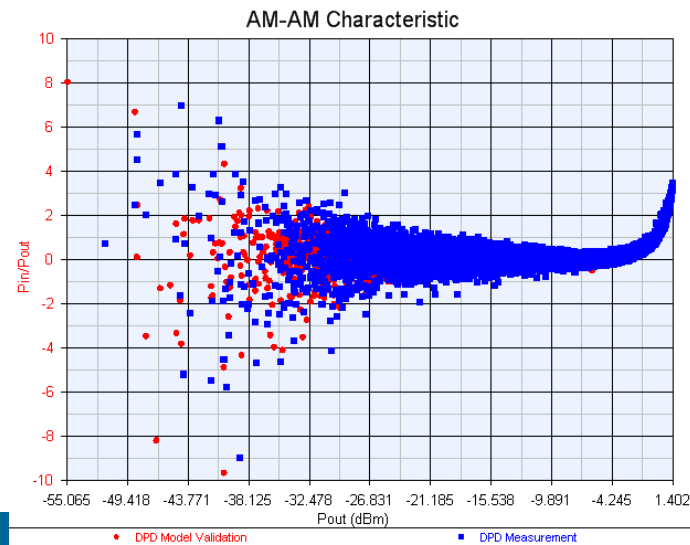


DPD Hardware Verification – LTE (Step 3)

Step 3: DUT Model Extraction



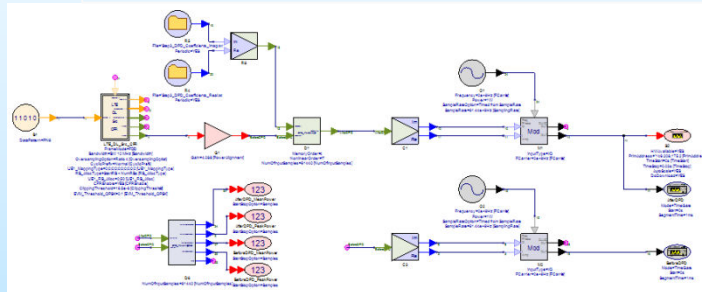
DPD Verification AM-AM



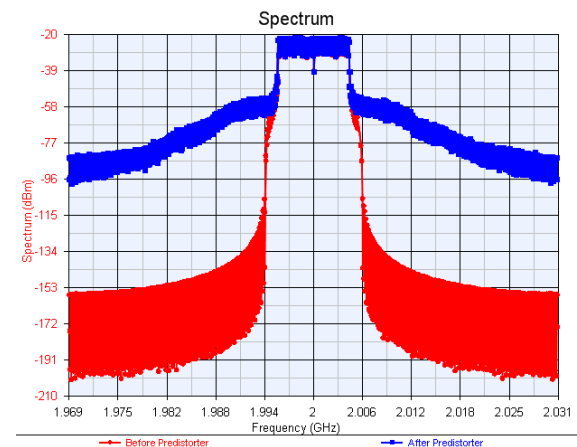
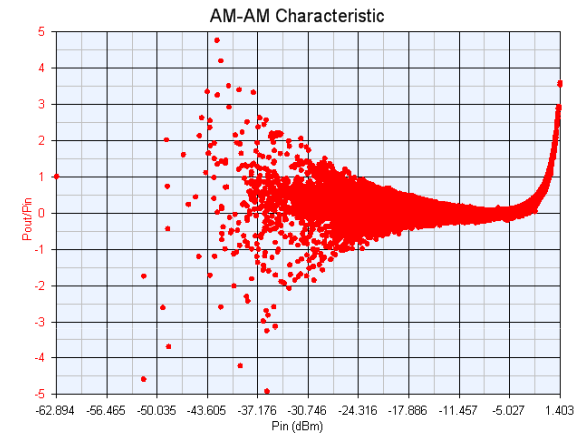
This step is to extract PA nonlinear coefficient from the PA input and PA output waveform and get the coefficients of the DPD model.

DPD Hardware Verification – LTE (Step 4)

Step 4: DUT Response



This step is to apply the DPD model extracted in Step 3. The generated LTE downlink signal is firstly pre-distorted by the extracted model, and then downloaded into the ESG.

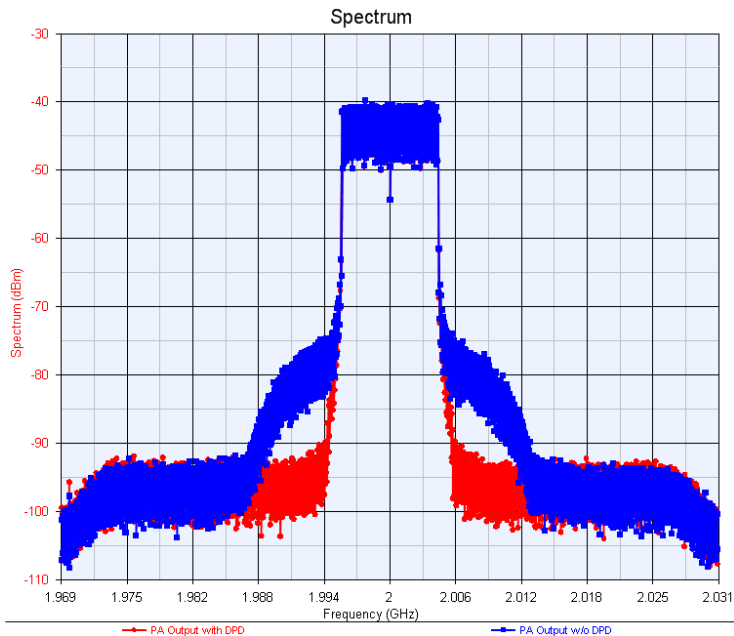


DPD Hardware Verification – LTE (Step 5)

Step 5: Verify DUT Response

Spectrum
EVM
ACLR

This step is to verify the performances of the DPD (including spectrums of the DUT output signal w/ and w/o DPD, EVM and ACLR).



Download Parameters

RFPower: -4.963 dBm
PrimAddress: 146.208.175.2
Time Start: 0 ms
Time Stop: 30 ms

LTE DL Source → MXG/ESG

This step is to download the original signal (without DPD) with the same RF power as step 4. You may also use the downloaded waveform in step 1 and change the power in MXG/ESG.

Download Waveform → Go To ESG Web Control

Capture Waveform

Show Results

Spectrum EVM ACLR

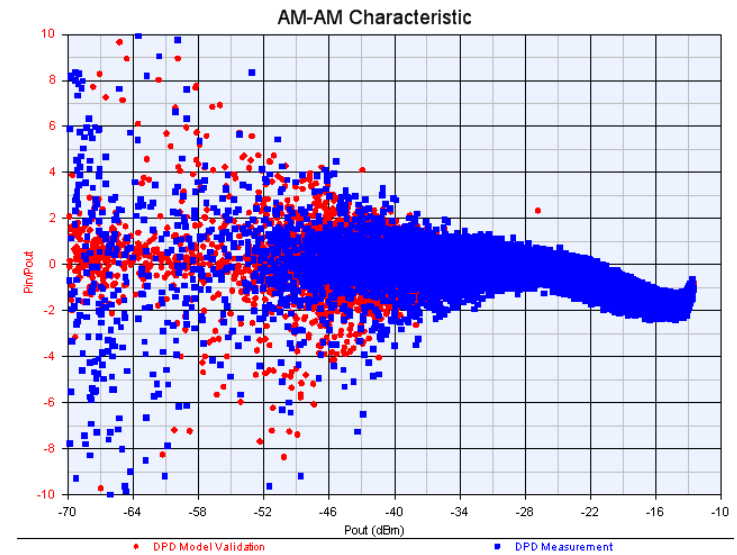
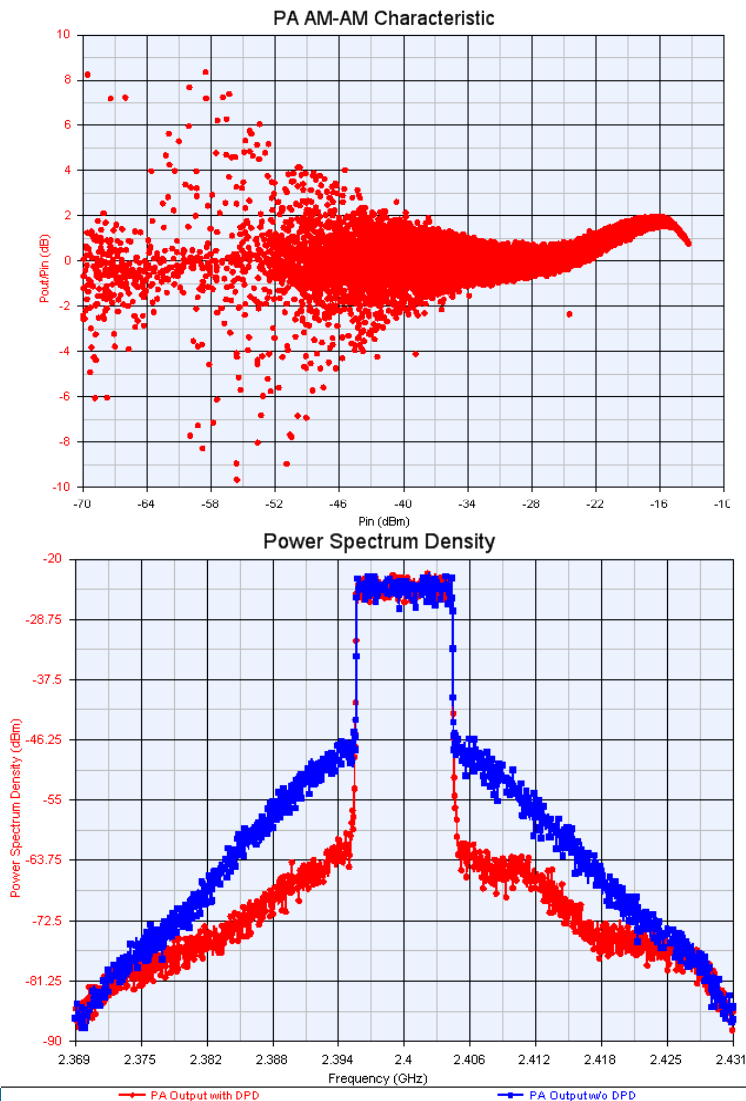
EVM (dB)

| Index | AfterDPD_PA | Original | After_PA |
|-------|-------------|----------|----------|
| 1 | -19.443 | -19.382 | -18.577 |
| 2 | -18.84 | -18.844 | -18.071 |

ACLR (dB)

| Original_ACLR_L_2BW | Original_ACLR_L_BW | Original_ACLR_U_BW | Original_ACLR_U_2BW |
|---------------------|--------------------|--------------------|---------------------|
| 53.179 | 50.653 | 51.252 | 53.363 |
| PA_ACLR_L_2BW | PA_ACLR_L_BW | PA_ACLR_U_BW | PA_ACLR_U_2BW |
| 51.307 | 38.244 | 38.361 | 51.23 |
| DPD_PA_ACLR_L_2BW | DPD_PA_ACLR_L_BW | DPD_PA_ACLR_U_BW | DPD_PA_ACLR_U_2BW |
| 52.437 | 50.514 | 50.611 | 52.825 |

Hardware Verification Results of Doherty PA



EVM (dB)

| Index | AfterDPD_PA | Original | After_PA |
|-------|-------------|----------|----------|
| 1 | -19.531 | -19.384 | -18.625 |
| 2 | -19.438 | -19.392 | -18.616 |

ACLR (dB)

| Original_ACLR_L_2BW | Original_ACLR_L_BW | Original_ACLR_U_BW | Original_ACLR_U_2BW |
|---------------------|--------------------|--------------------|---------------------|
| 51.614 | 50.992 | 50.005 | 51.821 |
| PA_ACLR_L_2BW | PA_ACLR_L_BW | PA_ACLR_U_BW | PA_ACLR_U_2BW |
| 43.585 | 27.643 | 27.422 | 42.262 |
| DPD_PA_ACLR_L_2BW | DPD_PA_ACLR_L_BW | DPD_PA_ACLR_U_BW | DPD_PA_ACLR_U_2BW |
| 50.147 | 42.426 | 42.082 | 47.779 |

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